



MAX PLANCK INSTITUTE
FOR DYNAMICS OF COMPLEX
TECHNICAL SYSTEMS
MAGDEBURG



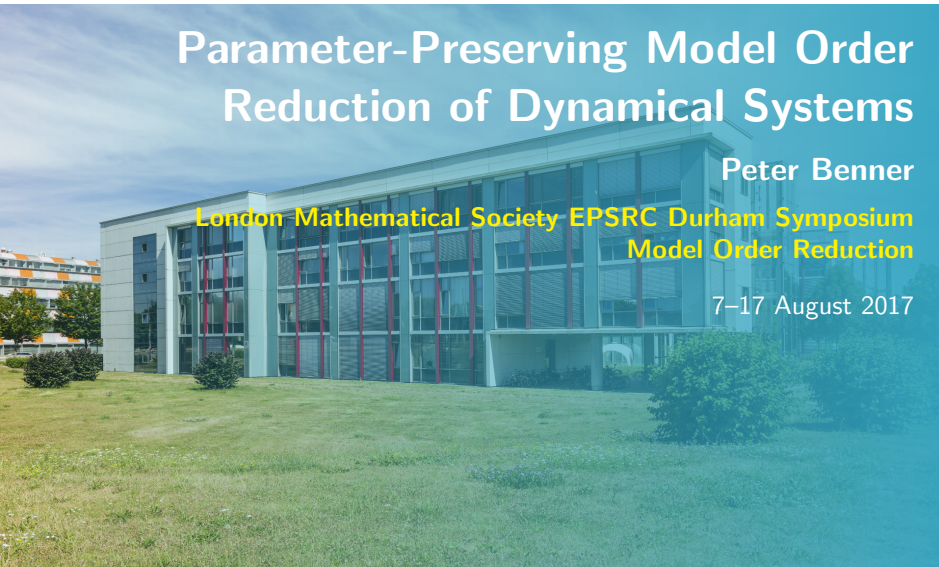
COMPUTATIONAL METHODS IN
SYSTEMS AND CONTROL THEORY

Parameter-Preserving Model Order Reduction of Dynamical Systems

Peter Benner

London Mathematical Society EPSRC Durham Symposium
Model Order Reduction

7–17 August 2017





MPI Magdeburg

- founded 1998
- 4 departments (directors)
- 10 research groups
- budget \sim 15 Mio. EUR
- \sim 230 employees
- 130 scientists,
- doing research in
 - biotechnology
 - chemical engineering
 - process engineering
 - energy conversion
 - applied math



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 P. Benner, S. Gugercin, and K. Willcox.

A Survey of Model Reduction Methods for Parametric Systems.
SIAM Review 57(4):483–531, 2015.



1. Introduction
2. PMOR Methods based on Interpolation
3. PMOR via Bilinearization
4. Conclusions and Outlook



1. Introduction

Parametric Dynamical Systems

Motivating Example

Parametric Modeling

The Parametric Model Order Reduction (PMOR) Problem

Error Measures

2. PMOR Methods based on Interpolation

3. PMOR via Bilinearization

4. Conclusions and Outlook



Parametric Dynamical Systems

$$\Sigma(p) : \begin{cases} E(p)\dot{x}(t; p) = f(t, x(t; p), u(t), p), & x(t_0) = x_0, & (a) \\ y(t; p) = g(t, x(t; p), u(t), p) & & (b) \end{cases}$$

with

- (generalized) **states** $x(t; p) \in \mathbb{R}^n$ ($E \in \mathbb{R}^{n \times n}$),
- **inputs (controls)** $u(t) \in \mathbb{R}^m$,
- **outputs (measurements, quantity of interest)** $y(t; p) \in \mathbb{R}^q$,
(b) is called **output equation**,
- $p \in \Omega \subset \mathbb{R}^d$ is a **parameter vector**, Ω is bounded.



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$E(p)$ singular \Rightarrow (a) is system of differential-algebraic equations (DAEs)
otherwise \Rightarrow (a) is system of ordinary differential equations (ODEs)



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Applications:

- Repeated simulation for varying material or geometry parameters, boundary conditions,
- control, optimization and design,
- of models, often generated by FE software (e.g., ANSYS, NASTRAN, ...) or automatic tools (e.g., Modelica).



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Underlying PDE and boundary conditions often not accessible!



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Underlying PDE and boundary conditions often not accessible!

Parametric discretized model often not available,
but matrices for certain parameter values can be extracted
(or output data for given u and p can be generated!)



Linear, Time-Invariant (Parametric) Systems

$$\begin{aligned} E(p)\dot{x}(t;p) &= A(p)x(t;p) + B(p)u(t), & A(p), E(p) &\in \mathbb{R}^{n \times n}, \\ y(t;p) &= C(p)x(t;p), & B(p) &\in \mathbb{R}^{n \times m}, C(p) \in \mathbb{R}^{q \times n}. \end{aligned}$$



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Laplace Transformation / Frequency Domain

Application of Laplace transformation

$$x(t; p) \mapsto x(s; p), \quad \dot{x}(t; p) \mapsto sx(s; p)$$

to linear system with $x(0; p) \equiv 0$:

$$sE(p)x(s; p) = A(p)x(s; p) + B(p)u(s), \quad y(s; p) = C(p)x(s; p),$$

yields I/O-relation in frequency domain:

$$y(s; p) = \underbrace{\left(C(p)(sE(p) - A(p))^{-1}B(p) \right)}_{=: G(s, p)} u(s).$$

$G(s, p)$ is the parameter-dependent **transfer function** of $\Sigma(p)$.



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Goal: **Fast evaluation** of mapping $(u, p) \rightarrow y(s; p)$.

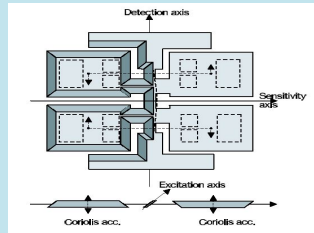


Microgyroscope (butterfly gyro)



- Voltage applied to electrodes induces vibration of wings, resulting rotation due to Coriolis force yields sensor data.
- FE model of second order:
 $N = 17.361 \rightsquigarrow n = 34.722, m = 1, q = 12.$
- Sensor for position control based on acceleration and rotation.

- Applications:
 - inertial navigation,
 - electronic stability control (ESP).

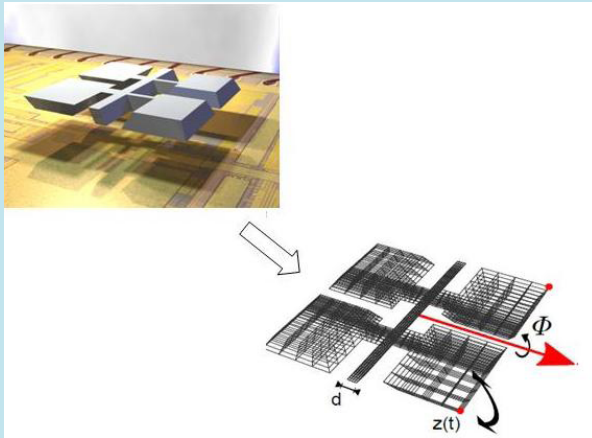


Source: MOR Wiki: <http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Gyroscope>



Microgyroscope (butterfly gyro)

Parametric FE model: $M(d)\ddot{x}(t) + D(\theta, d, \alpha, \beta)\dot{x}(t) + T(d)x(t) = Bu(t)$.





Microgyroscope (butterfly gyro)

Parametric FE model:

$$M(d)\ddot{x}(t) + D(\theta, d, \alpha, \beta)\dot{x}(t) + T(d)x(t) = Bu(t),$$

where

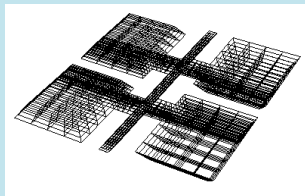
$$M(d) = M_1 + dM_2,$$

$$D(\theta, d, \alpha, \beta) = \theta(D_1 + dD_2) + \alpha M(d) + \beta T(d),$$

$$T(d) = T_1 + \frac{1}{d}T_2 + dT_3,$$

with the parameters

- width of bearing: d ,
- angular velocity: θ ,
- Rayleigh damping parameters: α, β .

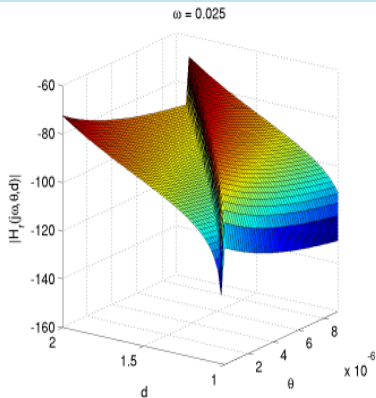
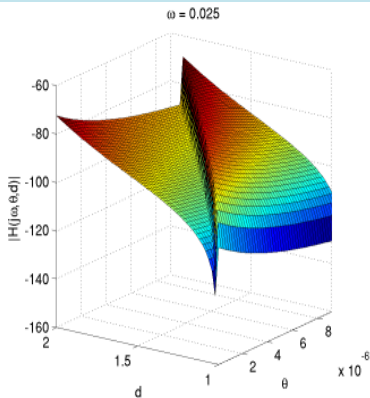




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Original...

and reduced-order model.





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Desirable: **parameter-affine** representation:

$$\begin{aligned} E(p) &= E_0 + e_1(p)E_1 + \dots + e_{q_E}(p)E_{q_E}, \\ A(p) &= A_0 + a_1(p)A_1 + \dots + a_{q_A}(p)A_{q_A}, \\ B(p) &= B_0 + b_1(p)B_1 + \dots + b_{q_B}(p)B_{q_B}, \\ C(p) &= C_0 + c_1(p)C_1 + \dots + c_{q_C}(p)C_{q_C}, \end{aligned}$$

allows easy parameter preservation for projection based model reduction.



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$$A(p) = A_0 + a_1(p)A_1 + \dots + a_{q_A}(p)A_{q_A}, \quad \dots$$

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W.l.o.g. may assume this affine representation:

- Any system can be written in this affine form for some $q_M \leq n^2$, but for efficiency, need $q_M \ll n!$ ($M \in \{E, A, B, C\}$)



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- Any system can be written in this affine form for some $q_M \leq n^2$, but for efficiency, need $q_M \ll n!$ ($M \in \{E, A, B, C\}$)
- Empirical (operator) interpolation [BARRAULT/MADAY/NGUYEN/PATERA '04] yields this structure for "smooth enough" parameter dependencies; variants (Q)DEIM et al.



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Obtaining a linear parametrization

Assumption: the parametric matrix $M(p)$ ($M \in \{A, B, C, E\}$) can be parameterized linearly:

$$M(p) = p_1 M_1 + \dots p_d M_d, \quad p \in \Omega \subset \mathbb{R}^d.$$



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Given: $M(p^{(j)})$ for $j = 1, \dots, d$. Then

$$p_1^{(j)} M_1 + p_2^{(j)} M_2 + \dots + p_d^{(j)} M_d = M(p^{(j)}), \quad j = 1, \dots, d,$$

is a linear system in the unknown matrices M_j , $j = 1, \dots, d$.



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Choice $p^{(j)} = e_j \in \mathbb{R}^d \rightsquigarrow M_j = M(p^{(j)})$, $j = 1, \dots, d$.

But e_j may not be feasible, i.e., $e_j \notin \Omega$, so that different physics would be modeled!



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With $P := \begin{bmatrix} p_i^{(j)} \end{bmatrix}_{i,j=1}^d \in \mathbb{R}^{d \times d}$,

$$(P \otimes I) \begin{bmatrix} M_1 \\ \vdots \\ M_d \end{bmatrix} = \begin{bmatrix} M(p^{(1)}) \\ \vdots \\ M(p^{(d)}) \end{bmatrix}.$$



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Now obtain unknown matrices as

$$\begin{bmatrix} M_1 \\ \vdots \\ M_d \end{bmatrix} = (P^{-1} \otimes I) \begin{bmatrix} M(p^{(1)}) \\ \vdots \\ M(p^{(d)}) \end{bmatrix}.$$



Obtaining a linear parametrization — Example

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$$M(p) = p_1 M_1 + \dots p_d M_d, \quad p \in \Omega \subset \mathbb{R}^d.$$

Parabolic system — linear diffusion-convection-reaction equation:

$$\partial_t \mathbf{x} - \kappa \Delta \mathbf{x} + [\nu_1, \nu_2] \cdot \nabla \mathbf{x} + \mu \mathbf{x} = \mathbf{b}(\xi) u(t), \quad \xi \in \mathcal{D} \subset \mathbb{R}^2, \quad t > 0$$

with initial/boundary conditions

$$\begin{aligned} \alpha(\xi) \mathbf{x} + \beta(\xi) \partial_\eta \mathbf{x} &= 0, & \xi \in \partial \mathcal{D}, \quad t \in [0, T], \\ \mathbf{x}(0, \xi) &= \mathbf{x}_0(\xi), & \xi \in \mathcal{D}, \end{aligned}$$



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FEM/FVM/FDM \rightsquigarrow

$$E \dot{\mathbf{x}}(t) = -(\kappa K + \nu_1 C_x + \nu_2 C_y + \mu R) \mathbf{x}(t) + B u(t),$$



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$$E \dot{\mathbf{x}}(t) = -(\kappa K + \nu_1 C_x + \nu_2 C_y + \mu R) \mathbf{x}(t) + B u(t),$$

Hence, we obtain a linear parametrization, only the "A" matrix is parameterized here:

$$\begin{aligned} A(p) &= p_1 A_1 + p_2 A_2 + p_3 A_3 + p_4 A_4, & \text{with} \\ p &= [\kappa, \nu_1, \nu_2, \mu] \quad \text{and} \quad A_1 = -K, \quad A_2 = -C_x, \quad A_3 = -C_y, \quad A_4 = -R. \end{aligned}$$

Note: selecting, e.g., $p^{(1)} = \mathbf{e}_1 \rightsquigarrow$ heat equation w/o convection/reaction \rightsquigarrow different physics!



Obtaining an affine parametrization using interpolation

If $M(p^{(j)})$ (for $j = 1, \dots, \ell$ and $M \in \{A, B, C, E\}$) are available (e.g., can be exported from FEM software), and parametrization is unknown anyway, using standard interpolation techniques to build parametric model:

Set $M_j := M(p^{(j)})$ and build for $M \in \{A, B, C, E\}$

$$M(p) = \phi_1(p)M_1 + \phi_2(p)M_2 + \dots + \phi_\ell(p)M_\ell, \quad p \in \Omega \subset \mathbb{R}^d,$$

where $\phi_k(p^{(j)}) = \delta_{kj}$, via

- polynomial interpolation (e.g., barycentric Lagrangian interpolation),
- rational (Loewner) interpolation,
- sparse grid interpolation, ...



Problem

Approximate the dynamical system

$$\begin{aligned} E(p)\dot{x} &= A(p)x + B(p)u, & E(p), A(p) &\in \mathbb{R}^{n \times n}, \\ y &= C(p)x, & B(p) &\in \mathbb{R}^{n \times m}, C(p) \in \mathbb{R}^{q \times n}, \end{aligned}$$

by reduced-order system

$$\begin{aligned} \hat{E}(p)\dot{\hat{x}} &= \hat{A}(p)\hat{x} + \hat{B}(p)u, & \hat{E}(p), \hat{A}(p) &\in \mathbb{R}^{r \times r}, \\ \hat{y} &= \hat{C}(p)\hat{x}, & \hat{B}(p) &\in \mathbb{R}^{r \times m}, \hat{C}(p) \in \mathbb{R}^{q \times r}, \end{aligned}$$

of order $r \ll n$, such that

$$\|y - \hat{y}\| = \|Gu - \hat{G}u\| \leq \|G - \hat{G}\| \cdot \|u\| < \text{tolerance} \cdot \|u\| \quad \forall p \in \Omega.$$



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\Rightarrow Approximation problem: $\min_{\text{order}(\hat{G}) \leq r} \|G - \hat{G}\|.$



- Approximate (for fast evaluation) function G , defined on $\mathbb{C} \times \Omega$.



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- But:

$$\begin{aligned} G : \mathbb{C} \times \Omega &\rightarrow \mathbb{C}^{q \times m}, \quad \Omega = [\alpha_1, \beta_1] \times \dots \times [\alpha_d, \beta_d], \\ G(s; p_1, \dots, p_d) &\in \mathbb{C}^{q \times m}. \end{aligned}$$

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Hence, try to preserve the structure of the original model!



Structure Preservation

Parametric System

$$\Sigma(p) : \begin{cases} E(p)\dot{x}(t; p) &= A(p)x(t; p) + B(p)u(t), \\ y(t; p) &= C(p)x(t; p). \end{cases}$$



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Parametric model reduction goal:

preserve parameters as *symbolic quantities* in reduced-order model:

$$\hat{\Sigma}(p) : \begin{cases} \hat{E}(p)\dot{\hat{x}}(t; p) &= \hat{A}(p)\hat{x}(t; p) + \hat{B}(p)u(t), \\ \hat{y}(t; p) &= \hat{C}(p)\hat{x}(t; p) \end{cases}$$

with states $\hat{x}(t; p) \in \mathbb{R}^r$ and $r \ll n$.



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Assuming **parameter-affine** representation:

$$\begin{aligned} E(p) &= E_0 + e_1(p)E_1 + \dots + e_{q_E}(p)E_{q_E}, \\ A(p) &= A_0 + a_1(p)A_1 + \dots + a_{q_A}(p)A_{q_A}, \\ B(p) &= B_0 + b_1(p)B_1 + \dots + b_{q_B}(p)B_{q_B}, \\ C(p) &= C_0 + c_1(p)C_1 + \dots + c_{q_C}(p)C_{q_C}, \end{aligned}$$

allows easy parameter preservation for projection based model reduction.



Structure Preservation

Petrov-Galerkin-type projection

For given projection matrices $V, W \in \mathbb{R}^{n \times r}$ with $W^T V = I_r$
($\rightsquigarrow (VW^T)^2 = VW^T$ is projector), compute

$$\hat{E}(p) = W^T E_0 V + e_1(p) W^T E_1 V + \dots + e_{q_E}(p) W^T E_{q_E} V$$

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Basis Generation — Global vs. Local

Local Bases

Obtain $V_k, W_k \in \mathbb{R}^{n \times r_k}$ using any non-parametric linear MOR method for a number of full-order models $\Sigma(p^{(k)})$, $k = 1, \dots, \ell$. Then compute reduced-order model by



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3. matrix interpolation: different models obtained in different coordinate systems \rightsquigarrow Procrustes problem \rightsquigarrow potential loss of accuracy; efficiency in "online" phase suffers from evaluating the interpolation operator.



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Avoids most of the problems encountered with local bases, but requires parameter-affine representation of system.

**Empirical Matrix Interpolation Method**

[B./GUGERCIN/WILCOX 2015]

Given $V, W \in \mathbb{R}^{n \times r}$ and suppose only that $M(p) \in \mathbb{R}^{n \times t}$ can be evaluated at specific parameter values.

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- Then $\hat{m}(p) = \text{vec}(\hat{M}(p)) \in \mathbb{R}^{rt}$ (or \mathbb{R}^{r^2} if $t = n$) can be computed cheaply and independent of n as

$$\begin{aligned}\hat{m}(p) &= \text{vec}\left(W^T M(p) V\right) \\ &= (V^T \otimes W^T) m(p) \approx (V^T \otimes W^T) \tilde{m}(p) = (V^T \otimes W^T) \Psi \alpha(p) = \tilde{\tilde{m}}(p).\end{aligned}$$



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- This is achieved by sampling $M(p)$ at $p = p^{(j)}$, $j = 1, \dots, \ell$, yielding

$$\psi_j = \text{vec}(M(p^{(j)})) \quad \text{and} \quad \Psi = [\psi_1, \dots, \psi_\ell].$$

Then apply (Q,D)EIM (or alike) to determine $\alpha(p)$ s.t. selected entries of $\tilde{m}(p)$ interpolate those entries of $m(p)$.



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Let z_1, z_2, \dots, z_ℓ be the selected indices to be exactly matched, and $Z := [e_{z_1}, \dots, e_{z_\ell}]$. Then, forcing interpolation at the selected rows implies

$$Z^T m(p) = Z^T \Psi \alpha(p) \implies \alpha(p) = (Z^T \Psi)^{-1} Z^T m(p).$$



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- Undoing the vectorization yields the reduced model matrix

$$\hat{M}(p) := \text{vec}^{-1}(\tilde{m}(p)) = \text{vec}^{-1}\left((V^T \otimes W^T) \Psi \alpha(p)\right) = \sum_{j=1}^{\ell} \alpha_j(p) \underbrace{W^T M(p^{(j)}) V}_{\text{precomputable!}}$$



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Parametric Systems Norms

Mean-square error norm:

$$\|G - \hat{G}\|_{\mathcal{H}_2 \otimes L_2(\Omega)}^2 := \frac{1}{2\pi} \int_{-\infty}^{+\infty} \int_{\Omega} \|G(j\omega, p) - \hat{G}(j\omega, p)\|_F^2 dp_1 \dots dp_d d\omega,$$

where $\|\cdot\|_F$ denotes the Frobenius norm.

Worst-case error norm:

$$\|G - \hat{G}\|_{\mathcal{H}_{\infty} \otimes L_{\infty}(\Omega)} := \sup_{\omega \in \mathbb{R}, p \in \Omega} \|G(j\omega, p) - \hat{G}(j\omega, p)\|_2.$$



1. Introduction

2. PMOR Methods based on Interpolation

- Interpolatory Model Reduction

- PMOR based on Multi-Moment Matching

- Optimal PMOR using Rational Interpolation?

- A Comparison of PMOR Methods

3. PMOR via Bilinearization

4. Conclusions and Outlook



Computation of reduced-order model by projection

Given a linear (descriptor) system $E\dot{x} = Ax + Bu, y = Cx$ with transfer function $G(s) = C(sE - A)^{-1}B$, a reduced-order model is obtained using truncation matrices $V, W \in \mathbb{R}^{n \times r}$ with $W^T V = I_r$ ($\leadsto (VW^T)^2 = VW^T$ is projector) by computing

$$\hat{E} = W^T E V, \hat{A} = W^T A V, \hat{B} = W^T B, \hat{C} = C V.$$

Petrov-Galerkin-type (two-sided) projection: $W \neq V$,

Galerkin-type (one-sided) projection: $W = V$.



Computation of reduced-order model by projection

Given a linear (descriptor) system $E\dot{x} = Ax + Bu$, $y = Cx$ with transfer function $G(s) = C(sE - A)^{-1}B$, a reduced-order model is obtained using truncation matrices $V, W \in \mathbb{R}^{n \times r}$ with $W^T V = I_r$ ($\leadsto (VW^T)^2 = VW^T$ is projector) by computing

$$\hat{E} = W^T E V, \quad \hat{A} = W^T A V, \quad \hat{B} = W^T B, \quad \hat{C} = C V.$$

Petrov-Galerkin-type (two-sided) projection: $W \neq V$,

Galerkin-type (one-sided) projection: $W = V$.

Rational Interpolation/Moment-Matching

Choose V, W such that

$$G(s_j) = \hat{G}(s_j), \quad j = 1, \dots, k,$$

and

$$\frac{d^i}{ds^i} G(s_j) = \frac{d^i}{ds^i} \hat{G}(s_j), \quad i = 1, \dots, K_j, \quad j = 1, \dots, k.$$

**Theorem (simplified)** [GRIMME 1997, VILLEMAGNE/SKELTON 1987]

If

$$\begin{aligned}\text{span} \{ (s_1 E - A)^{-1} B, \dots, (s_k E - A)^{-1} B \} &\subset \text{range}(V), \\ \text{span} \{ (s_1 E - A)^{-T} C^T, \dots, (s_k E - A)^{-T} C^T \} &\subset \text{range}(W),\end{aligned}$$

then

$$G(s_j) = \hat{G}(s_j), \quad \frac{d}{ds} G(s_j) = \frac{d}{ds} \hat{G}(s_j), \quad \text{for } j = 1, \dots, k.$$

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Remarks:computation of V, W from **rational Krylov subspaces**, e.g.,

- dual rational Arnoldi/Lanczos [GRIMME 1997],
- **Iter. Rational Krylov-Alg. (IRKA)** [ANTOULAS/BEATTIE/GUGERCIN 2006/08].

**Theorem (simplified)** [GRIMME 1997, VILLEMAGNE/SKELTON 1987]

If

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Remarks:

using Galerkin/one-sided projection ($W \equiv V$) yields $G(s_j) = \hat{G}(s_j)$, but in general

$$\frac{d}{ds} G(s_j) \neq \frac{d}{ds} \hat{G}(s_j).$$

**Theorem (simplified)** [GRIMME 1997, VILLEMAGNE/SKELTON 1987]

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Remarks:

 $k = 1$, standard Krylov subspace(**s**) of dimension K :

$$\text{range}(V) = \mathcal{K}_K((s_1 E - A)^{-1}, (s_1 E - A)^{-1} B).$$

 \rightsquigarrow **moment-matching methods**/Padé approximation,

$$\frac{d^i}{ds^i} G(s_1) = \frac{d^i}{ds^i} \hat{G}(s_1), \quad i = 0, \dots, K-1(+K).$$



Numerical Example: A Printed Circuit Board (PCB)

- System in time domain:

$$\begin{aligned} E\dot{x}(t) &= Ax(t) + Bu(t), \\ y(t) &= Cx(t). \end{aligned}$$

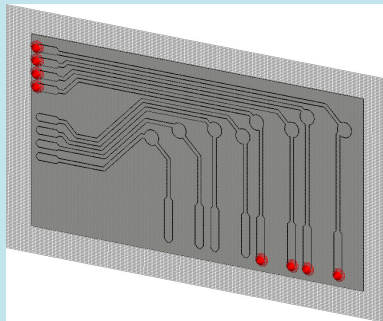
- System in frequency domain:

$$\begin{aligned} sEx(s) &= Ax(s) + Bu(s), \\ y(s) &= Cx(s). \end{aligned}$$

- Reduced basis method considers s as a parameter, and uses the system in frequency domain to compute $\text{range}(V) = \text{span}\{x(s_1), \dots, x(s_m)\}$.

The ROM is obtained by Galerkin projection with V .

Printed circuit board



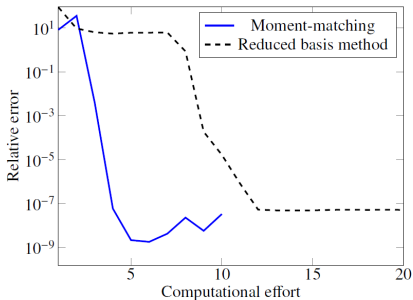
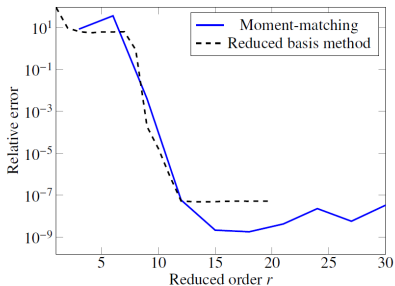
$$n = 233,060, m = q = 1.$$

Courtesy of TEMF, TU Darmstadt.



Numerical Example: A Printed Circuit Board (PCB)

Moment-matching vs. reduced basis method





\mathcal{H}_2 -Model Reduction for Linear Systems

Consider **stable** (i.e. $\Lambda(A) \subset \mathbb{C}^-$) linear systems Σ ,

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) \quad \simeq \quad y(s) = \underbrace{C(sI - A)^{-1}B}_{=: G(s)} u(s)$$

System norms

Recall: two common system norms for measuring approximation quality are

- \mathcal{H}_2 -norm, $\|\Sigma\|_{\mathcal{H}_2} = \left(\frac{1}{2\pi} \int_0^{2\pi} \text{tr} \left((G^T(-j\omega)G(j\omega)) \right) d\omega \right)^{\frac{1}{2}},$
- \mathcal{H}_∞ -norm, $\|\Sigma\|_{\mathcal{H}_\infty} = \sup_{\omega \in \mathbb{R}} \sigma_{\max}(G(j\omega)),$

where

$$G(s) = C(sI - A)^{-1}B.$$

Error system and \mathcal{H}_2 -Optimality

[MEIER/LUENBERGER 1967]

In order to find an \mathcal{H}_2 -optimal reduced system, consider the **error system** $G(s) - \hat{G}(s)$ which can be realized by

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = [C \quad -\hat{C}].$$

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Assuming a coordinate system in which \hat{A} is diagonal and taking derivatives of

$$\|G(\cdot) - \hat{G}(\cdot)\|_{\mathcal{H}_2}^2$$

with respect to free parameters in $\Lambda(\hat{A}), \hat{B}, \hat{C} \rightsquigarrow$ **first-order necessary \mathcal{H}_2 -optimality conditions (SISO)**

$$G(-\hat{\lambda}_i) = \hat{G}(-\hat{\lambda}_i),$$

$$G'(-\hat{\lambda}_i) = \hat{G}'(-\hat{\lambda}_i),$$

where $\hat{\lambda}_i$ are the poles of the reduced system $\hat{\Sigma}$.

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First-order necessary \mathcal{H}_2 -optimality conditions (MIMO):

$$\begin{aligned} G(-\hat{\lambda}_i) \tilde{B}_i &= \hat{G}(-\hat{\lambda}_i) \tilde{B}_i, & \text{for } i = 1, \dots, r, \\ \tilde{C}_i^T G(-\hat{\lambda}_i) &= \tilde{C}_i^T \hat{G}(-\hat{\lambda}_i), & \text{for } i = 1, \dots, r, \\ \tilde{C}_i^T H'(-\hat{\lambda}_i) \tilde{B}_i &= \tilde{C}_i^T \hat{G}'(-\hat{\lambda}_i) \tilde{B}_i, & \text{for } i = 1, \dots, r, \end{aligned}$$

where $\hat{A} = R \hat{\Lambda} R^{-T}$ is the spectral decomposition of the reduced system and $\tilde{B} = \hat{B}^T R^{-T}$, $\tilde{C} = \hat{C} R$.

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$$\begin{aligned} \Leftrightarrow \operatorname{vec}(I_q)^T \left(e_j e_i^T \otimes C \right) \left(-\hat{\Lambda} \otimes I_n - I_r \otimes A \right)^{-1} \left(\tilde{B}^T \otimes B \right) \operatorname{vec}(I_m) \\ = \operatorname{vec}(I_q)^T \left(e_j e_i^T \otimes \hat{C} \right) \left(-\hat{\Lambda} \otimes I_r - I_r \otimes \hat{A} \right)^{-1} \left(\tilde{B}^T \otimes \hat{B} \right) \operatorname{vec}(I_m), \\ \text{for } i = 1, \dots, r \text{ and } j = 1, \dots, q. \end{aligned}$$



Interpolation of the Transfer Function [GRIMME 1997]

Construct reduced transfer function by **Petrov-Galerkin** projection $\mathcal{P} = VW^T$, i.e.

$$\hat{G}(s) = CV (sI - W^T A V)^{-1} W^T B,$$



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where V and W are given as

$$V = [(-\mu_1 I - A)^{-1} B, \dots, (-\mu_r I - A)^{-1} B],$$
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for $i = 1, \dots, r$.

Starting with an initial guess for $\hat{\Lambda}$ and setting $\mu_i \equiv \hat{\lambda}_i \rightsquigarrow$ iterative algorithms (IRKA/MIRIAM) that yield \mathcal{H}_2 -optimal models.

[GUGERCIN ET AL. 2006/08], [BUNSE-GERSTNER ET AL. 2007],
[VAN DOOREN ET AL. 2008]



The Basic IRKA Algorithm

Algorithm 1 IRKA (MIMO version/MIRIAM)

Input: A stable, B , C , \hat{A} stable, \hat{B} , \hat{C} , $\delta > 0$.

Output: A^{opt} , B^{opt} , C^{opt}

- 1: **while** $(\max_{j=1,\dots,r} \left\{ \frac{|\mu_j - \mu_j^{\text{old}}|}{|\mu_j|} \right\} > \delta)$ **do**
- 2: $\text{diag}(\mu_1, \dots, \mu_r) := R^{-1} \hat{A} R = \text{spectral decomposition.}$
- 3: $\tilde{B} = \hat{B}^H R^{-T}$, $\tilde{C} = \hat{C} R$.
- 4: $V = \left[(-\mu_1 I - A)^{-1} B \tilde{b}_1, \dots, (-\mu_r I - A)^{-1} B \tilde{b}_r \right]$
- 5: $W = \left[(-\mu_1 I - A^T)^{-1} C^T \tilde{c}_1, \dots, (-\mu_r I - A^T)^{-1} C^T \tilde{c}_r \right]$
- 6: $V = \text{orth}(V)$, $W = \text{orth}(W)$, $W = W(V^H W)^{-1}$
- 7: $\hat{A} = W^H A V$, $\hat{B} = W^H B$, $\hat{C} = C V$.
- 8: **end while**
- 9: $A^{opt} = \hat{A}$, $B^{opt} = \hat{B}$, $C^{opt} = \hat{C}$.



Idea: choose appropriate frequency parameter \hat{s} and parameter vector \hat{p} , expand into multivariate power series about (\hat{s}, \hat{p}) and compute reduced-order model, so that

$$G(s, p) = \hat{G}(s, p) + \mathcal{O}\left(|s - \hat{s}|^K + \|p - \hat{p}\|^L + |s - \hat{s}|^k \|p - \hat{p}\|^l\right),$$

i.e., first $K, L, k + l$ (mostly: $K = L = k + l$) coefficients (**multi-moments**) of Taylor/Laurent series coincide.



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Algorithms:

- [1] [DANIEL ET AL. 2004]: explicit computation of moments, numerically unstable.
- [2] [FARLE ET AL. 2006/07]: Krylov subspace approach, only polynomial param.-dependance, numerical properties not clear, but appears to be robust.
- [3] [WEILE ET AL. 1999, FENG/B. 2007/14]: Arnoldi-MGS method, employ recursive dependance of multi-moments, numerically robust, r often larger as for [2].
- [4] **New:** employ dual-weighted residual error bound and greedy procedure to define interpolation points an $\#$ of multi-moments matched [ANTOULAS/B./FENG 2014/17].



Parametric System

Again, consider linear parametric system

$$\Sigma(p) : \begin{cases} E(p)\dot{x}(t; p) &= A(p)x(t; p) + B(p)u(t), \\ y(t; p) &= C(p)x(t; p) \end{cases}$$

together with its transfer function $G(s, p)$.



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For simplicity, assume $B(\mu) \equiv B$, and re-parameterize — $\mu := [s, p^T, \dots]^T \in \mathbb{C}^\ell$ such that with

$$G(\mu) \equiv G(s, p), \quad x(\mu) \equiv x(s, p), \quad y(\mu) \equiv y(s, p), \dots$$
$$\mathcal{A}(\mu) := sE(p) - A(p),$$

we obtain linear-affine structure of $\mathcal{A}(\mu)$:

$$\mathcal{A}(\mu) = \mathcal{A}_0 + \mu_1 \mathcal{A}_1 + \dots + \mu_\ell \mathcal{A}_\ell.$$



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In frequency domain, we may then re-write the parametric system as

$$\mathcal{A}(\mu)x(\mu) = Bu(s), \quad y(\mu) = C(\mu)x(\mu).$$



Multivariate Power Series Expansion I

Choose an expansion point $\mu^{(0)}$, and write

$$\begin{aligned}\mathcal{A}(\mu) &= \mathcal{A}_0 + \mu_1 \mathcal{A}_1 + \dots + \mu_\ell \mathcal{A}_m \\ &= \underbrace{(\mathcal{A}_0 + \mu_1^{(0)} \mathcal{A}_1 + \dots + \mu_\ell^{(0)} \mathcal{A}_m)}_{:= \mathcal{M}_0} + \left((\mu_1 - \mu_1^{(0)}) \mathcal{A}_1 + \dots + (\mu_\ell - \mu_\ell^{(0)}) \mathcal{A}_\ell \right)\end{aligned}$$



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Using the **Neumann lemma** ($(I - F)^{-1} = \sum_{j=0}^{\infty} F^j$ if $\|F\| < 1$), we obtain

$$\begin{aligned}\mathcal{A}(\mu)^{-1} &= \sum_{j=0}^{\infty} (-1)^j \left((\mu_1 - \mu_1^{(0)}) \mathcal{M}_0^{-1} \mathcal{A}_1 + \dots + (\mu_\ell - \mu_\ell^{(0)}) \mathcal{M}_0^{-1} \mathcal{A}_\ell \right)^j \mathcal{M}_0^{-1} \\ &= \sum_{j=0}^{\infty} (\sigma_1 \mathcal{M}_1 + \dots + \sigma_\ell \mathcal{M}_\ell)^j \mathcal{M}_0^{-1},\end{aligned}$$

where $\sigma_i = \mu_i - \mu_i^{(0)}$ and $\mathcal{M}_i = -\mathcal{M}_0^{-1} \mathcal{A}_i$ for $i = 1, \dots, \ell$.



Multivariate Power Series Expansion II

We have

$$\mathcal{A}(\mu)x(\mu) = Bu(s).$$

and

$$\mathcal{A}(\mu)^{-1} = \sum_{j=0}^{\infty} (\sigma_1 \mathcal{M}_1 + \dots \sigma_\ell \mathcal{M}_\ell)^j \mathcal{M}_0^{-1},$$

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Hence,

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Thus, $x(\mu)$ is (approximately, locally) contained in the Krylov subspace $\mathcal{K}_{k+1}((\sigma_1 \mathcal{M}_1 + \dots + \sigma_\ell \mathcal{M}_\ell), \mathcal{B})$.



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- Project the state-space onto this subspace.



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- The ROM is obtained by Galerkin projection.
- Petrov-Galerkin projection possible using the "dual" Krylov subspace obtained from using \mathcal{A}^T and C^T [AHMAD/B./FENG 2017].



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- Project the state-space onto this subspace.
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- Approximation is only valid locally (convergence radius of Neumann series!)
 \rightsquigarrow use several expansion points $\mu^{(0)}, \dots, \mu^{(h)}$, and concatenate (and truncate) the local bases to obtain a global basis.



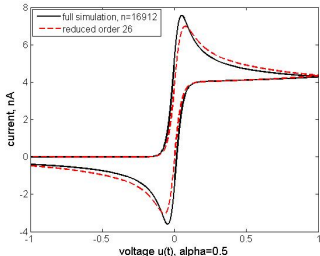
Numerical Examples: Electro-Chemical SEM

Compute cyclic voltammogram based on FE model

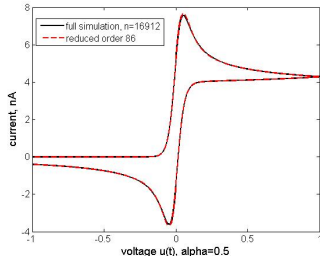
$$E\dot{x}(t) = (A_0 + p_1 A_1 + p_2 A_2)x(t) + Bu(t), \quad y(t) = c^T x(t),$$

where $n = 16,912$, $m = 3$, A_1, A_2 diagonal.

$$K = L = k + \ell = 4 \Rightarrow r = 26$$



$$K = L = k + \ell = 9 \Rightarrow r = 86$$



Source: MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Scanning_Electrochemical_Microscopy



Open question

How to adaptively choose $\mu^{(i)}$?



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And how many partial derivatives to be matched at each interpolation point?



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Possible approach: adopt ideas from Reduced Basis Methods, i.e., let

$$\|G(\mu) - \hat{G}(\mu)\| \leq \Delta(\mu) \quad \text{or} \quad \|y(\mu) - \hat{y}(\mu)\| \leq \Delta_o(\mu)$$

guide the selection of $\mu^{(i)}$ for computable *a posteriori* error bounds for the state or the output.



Theorem (SISO case)

[FENG/ANTOULAS/B. 2015/17]

Assume that $\sigma_{\min}(G(s, p)) =: \beta(s, p) > 0 \quad \forall \operatorname{Re}(s) \geq 0, \forall p \in \Omega$, then

$$|H(s, p) - \hat{H}(s, p)| \leq \tilde{\Delta}(s, p) + |(\hat{x}^{du})^H \mathbf{r}^{pr}(s, p)| =: \Delta(s, p),$$

where

$$\tilde{\Delta}(s, p) = \frac{\|r^{du}(s, p)\|_2 \|r^{pr}(s, p)\|_2}{\beta(s, p)},$$

with the primal and dual residuals r^{pr}, r^{du} and the reduced "dual state" \hat{x}^{du} :

$$r^{pr}(s, p) = \|(B - (sE(p) - A(p))) (V(s\hat{E}(p) - \hat{A}(p))^{-1} \hat{B})\|,$$

$$r^{du}(s, p) = \|(C^T - (\bar{s}E(p) - A(p))^T) \hat{x}^{du}\|,$$

$$\hat{x}^{du} = -V^{du}(\bar{s}\hat{E}^{du}(p) - \hat{A}^{du}(p))^{-T} \hat{C}^{du}.$$

The dual reduced-order system is computed using Galerkin projection with V^{du} obtained by applying multi-moment matching algorithm to "dual" system $(\bar{s}E(p)^T - A(p)^T, C^T)$.



Remarks

- For application in "RBM fashion", $r^{du}(\mu)$, $r^{pr}(\mu)$ can be efficiently computed, need to solve sparse linear systems on training set, i.e., one sparse factorization for each sampling point.
- $\beta(s, p) = \sigma_{\min}(G(s, p))$ easily computable on the training set — system solves for evaluation of the transfer function readily available from residual computation!
- Extension to MIMO case possible taking max over all I/O channels.
- Can use Petrov-Galerkin framework using $W = V^{du}$ at no extra cost!



Algorithm 2 Automatic generation of the ROM: adaptively selecting $\mu^{(i)}$

Input: $V = []$; $\epsilon > \epsilon_{tol}$; Initial expansion point: $\hat{\mu}$; $i := -1$;
 Ξ_{train} : a set of samples of μ covering the parameter domain.

Output: V .

```
1: while  $\epsilon > \epsilon_{tol}$  do  
2:    $i = i + 1$ ;  
3:    $\mu^{(i)} = \hat{\mu}$ ;  
4:    $V_{\mu^{(i)}} = \text{orthogonal basis of } \mathcal{K}_{k+1}((\sigma_1^{(i)} \mathcal{M}_1 + \dots + \sigma_\ell^{(i)} \mathcal{M}_\ell), \mathcal{B})$ ;  
5:    $V = \text{orth}([V, V_{\mu^{(i)}}])$ ;  
6:    $\hat{\mu} = \arg \max_{\mu \in \Xi_{\text{train}}} \Delta(\mu)$ ;  
7:    $\epsilon = \Delta(\hat{\mu})$ ;  
8: end while
```



Numerical Example: Silicon Nitride Membrane

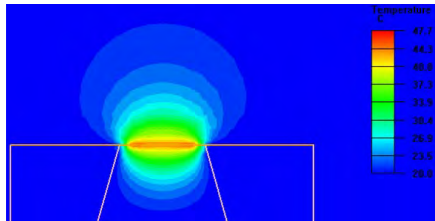
A SiN membrane can be a part of a gas sensor, an infra-red sensor, a microthruster, etc. Heat transfer in the membrane is described by

$$\begin{aligned}(E_0 + \rho c_p E_1) \dot{x}(t) &= -(K_0 + \kappa K_1 + h K_2) x(t) + b u(t) \\ y(t) &= C x(t),\end{aligned}$$

with parameters

- density $\rho \in [3000, 3200]$,
- specific heat capacity $c_p \in [400, 750]$,
- thermal conductivity $\kappa \in [2.5, 4]$,
- membrane heat transfer coefficient $h \in [10, 12]$.

and frequency $f \in [0, 25] \text{ Hz}$.



Source: MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Silicon_nitride_membrane



Numerical Examples: Silicon Nitride Membrane

Setting

- Training set: $\Xi_{train} = 5$ random samples for ρ and c_p , 3 random samples for κ and h , respectively, 10 samples of Laplace variable s .

- Error measures:

$$\varepsilon_{true}^{rel} = \max_{\mu \in \Xi_{train}} |G(\mu) - \hat{G}(\mu)| / |G(\mu)|,$$

$$\Delta^{rel}(\mu) = \Delta(\mu) / |\hat{G}(\mu)|$$

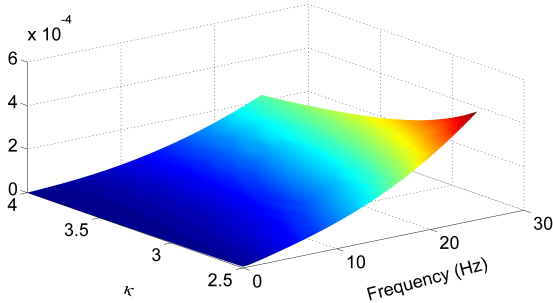
- $V_{\mu^{(i)}} = \text{span}\{\mathcal{B}, (\sigma_1^{(i)} \mathcal{M}_1 + \dots + \sigma_\ell^{(i)} \mathcal{M}_\ell) \mathcal{B}\}$, $\varepsilon_{tol}^{re} = 10^{-2}$, $n = 60,020$, $r = 8$.

iter.	ε_{true}^{rel}	$\Delta^{rel}(\mu^{(i)})$	s	ρc_p	κ	h
1	1×10^{-3}	3.44	18.94	1.37×10^6	2.74	10.97
2	1×10^{-4}	4.59×10^{-2}	0.89	1.31×10^6	3.96	11.60
3	2.80×10^{-5}	4.07×10^{-2}	23.98	2.35×10^6	3.94	10.28
4	2.58×10^{-6}	2.62×10^{-5}	0.89	2.31×10^6	2.74	10.28



Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for κ over set Ξ_{fine} with 16 equidistant samples of κ , 51 equidistant samples of the frequency f , while the other parameters are fixed.

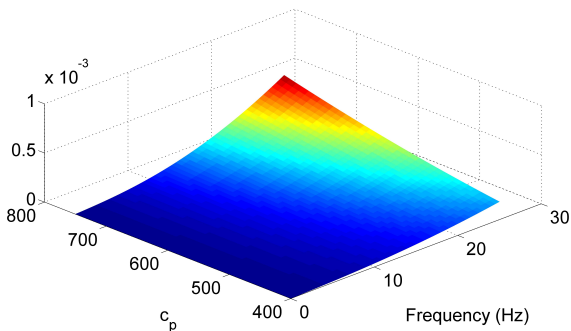


Relative error of the final ROM changing with κ and frequency.



Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for c_p over set Ξ_{fine} with 36 equidistant samples of c_p , 51 equidistant samples of the frequency f , while the other parameters are fixed.

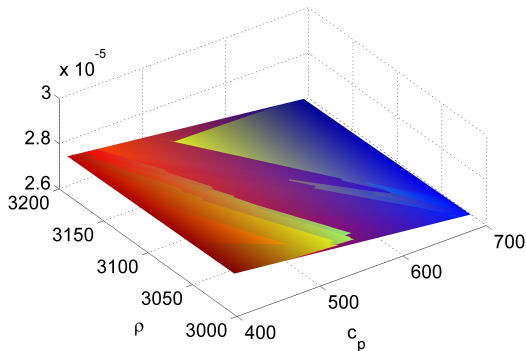


Relative error of the final ROM changing with c_p and frequency.



Numerical Examples: Silicon Nitride Membrane

Verification of the accuracy of the ROM for ρ , c_p over set Ξ_{fine} with 50 random samples of ρ , c_p , respectively, the other parameters are fixed.



Relative error of the final ROM changing with c_p and κ .



Greedy expansion point selection has a heuristic nature and relies on a training set.



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How to determine the right number of partial derivatives to be matched at the expansion points is an open problem (for potential solutions in the non-parametric case, see [FENG/KORVINK/B. 2015, BONIN/FASSBENDER/SOPPA/ZÄH 2016, LEE/CHU/FENG 2006,...]).



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Following the non-parametric case, one would need:

- Projection-based framework for tangential rational interpolation. [✓]



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Following the non-parametric case, one would need:

- Projection-based framework for tangential rational interpolation. [✓]
- Iterative procedure for selecting interpolation points. [x] ... [✓] for special case.



Theory: Interpolation of the Transfer Function

Theorem

[BAUR/BEATTIE/B./GUGERCIN 2007/11]

Let

$$\begin{aligned}\hat{G}(s, p) &:= \hat{C}(p)(s\hat{E}(p) - \hat{A}(p))^{-1}\hat{B}(p) \\ &= C(p)V(sW^T E(p)V - W^T A(p)V)^{-1}W^T B(p).\end{aligned}$$

Suppose $\hat{p} = [\hat{p}_1, \dots, \hat{p}_d]^T$ and $\hat{s} \in \mathbb{C}$ are chosen such that both $\hat{s} E(\hat{p}) - A(\hat{p})$ and $\hat{s} \hat{E}(\hat{p}) - \hat{A}(\hat{p})$ are invertible. If

$$(\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} B(\hat{p}) \in \text{range}(V)$$

or

$$\left(C(\hat{p}) (\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} \right)^T \in \text{range}(W),$$

then $G(\hat{s}, \hat{p}) = \hat{G}(\hat{s}, \hat{p})$.



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Extension to MIMO case using **tangential interpolation**: let $0 \neq b \in \mathbb{R}^m$, $0 \neq c \in \mathbb{R}^q$.

- a) If $(\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} B(\hat{p})b \in \text{range}(V)$, then $G(\hat{s}, \hat{p})b = \hat{G}(\hat{s}, \hat{p})b$.
- b) If $\left(c^T C(\hat{p}) (\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} \right)^T \in \text{range}(W)$, then $c^T G(\hat{s}, \hat{p}) = c^T \hat{G}(\hat{s}, \hat{p})$.



Theory: Interpolation of the Parameter Gradient

Theorem

[BAUR/BEATTIE/B./GUGERCIN 2007/11]

Suppose that $E(p)$, $A(p)$, $B(p)$, $C(p)$ are C^1 in a neighborhood of $\hat{p} = [\hat{p}_1, \dots, \hat{p}_d]^T$ and that both $\hat{s} E(\hat{p}) - A(\hat{p})$ and $\hat{s} \hat{E}(\hat{p}) - \hat{A}(\hat{p})$ are invertible.

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Note: result extends to MIMO case using **tangential interpolation**:

Let $0 \neq b \in \mathbb{R}^m$, $0 \neq c \in \mathbb{R}^q$ be arbitrary. If $(\hat{s} E(\hat{p}) - A(\hat{p}))^{-1} B(\hat{p})b \in \text{range}(V)$ and $(c^T C(\hat{p}) (\hat{s} E(\hat{p}) - A(\hat{p}))^{-1})^T \in \text{range}(W)$, then $\nabla_p c^T G(\hat{s}, \hat{p})b = \nabla_p c^T \hat{G}(\hat{s}, \hat{p})b$.



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1. Reduced-order model satisfies necessary conditions for surrogate models in trust region methods [ALEXANDROV/DENNIS/LEWIS/TORCZON 1998].
2. Approximation of gradient allows use of reduced-order model for sensitivity analysis.



Generic implementation of interpolatory PMOR

Define $\mathcal{A}(s, p) := sE(p) - A(p)$.

1. Select “frequencies” $s_1, \dots, s_k \in \mathbb{C}$ and parameter vectors $p^{(1)}, \dots, p^{(\ell)} \in \Omega \subset \mathbb{R}^d$.
2. Compute (orthonormal) basis of

$$\mathcal{V} = \text{span} \{ \mathcal{A}(s_1, p^{(1)})^{-1} B(p^{(1)}), \dots, \mathcal{A}(s_k, p^{(\ell)})^{-1} B(p^{(\ell)}) \}.$$

3. Compute (orthonormal) basis of

$$\mathcal{W} = \text{span} \{ \mathcal{A}(s_1, p^{(1)})^{-T} C(p^{(1)})^T, \dots, \mathcal{A}(s_k, p^{(\ell)})^{-T} C(p^{(\ell)})^T \}.$$

4. Set $V := [v_1, \dots, v_{k\ell}]$, $\tilde{W} := [w_1, \dots, w_{k\ell}]$, and $W := \tilde{W}(\tilde{W}^T V)^{-1}$.
(Note: $r = k\ell$).

5. Compute
$$\begin{cases} \hat{A}(p) := W^T A(p) V, & \hat{B}(p) := W^T B(p) V, \\ \hat{C}(p) := W^T C(p) V, & \hat{E}(p) := W^T E(p) V. \end{cases}$$



Remarks

- If directional derivatives w.r.t. p are included in $\text{range}(V)$, $\text{range}(W)$, then also the Hessian of $G(\hat{s}, \hat{p})$ is interpolated by the Hessian of $\hat{G}(\hat{s}, \hat{p})$.



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$$\|G(\cdot, p^{(k)}) - \hat{G}_*^{(k)}(\cdot)\|_{\mathcal{H}_2} = \min_{\substack{\text{order}(\hat{G})=r_k \\ \hat{G} \text{ stable}}} \|G(\cdot, p^{(k)}) - \hat{G}^{(k)}(\cdot)\|_{\mathcal{H}_2},$$

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- Optimal choice of interpolation **frequencies** s_k and **parameter vectors** $p^{(k)}$ possible for special cases.



Numerical Example: Thermal Conduction in a Semiconductor Chip

- Important requirement for a compact model of thermal conduction is boundary condition independence.
- The thermal problem is modeled by the heat equation, where heat exchange through device interfaces is modeled by convection boundary conditions containing **film coefficients** $\{p_i\}_{i=1}^3$ describing the heat exchange at i th interface.
- Spatial semi-discretization leads to

$$E\dot{x}(t) = (A_0 + \sum_{i=1}^3 p_i A_i)x(t) + bu(t), \quad y(t) = c^T x(t),$$

where $n = 4,257$, A_i , $i = 1, 2, 3$, are diagonal.

Source: C.J.M Lasance, *Two benchmarks to facilitate the study of compact thermal modeling phenomena*, IEEE Transactions on Components and Packaging Technologies, 24(4):559–565, 2001.

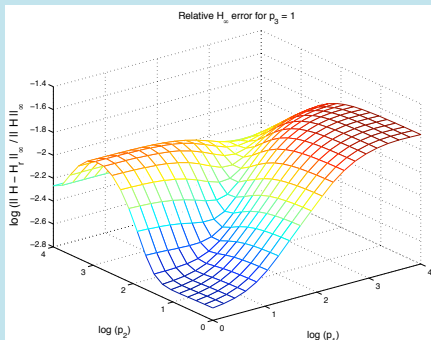
MOR Wiki: http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Microthruster_Unit



Numerical Example: Thermal Conduction in a Semiconductor Chip

Choose 2 interpolation points for parameters (“important” configurations), $8/7$ H_2 -optimal interpolation frequencies selected by IRKA. $\Rightarrow k = 2, \ell = 8, 7$, hence $r = 15$.

$p_3 = 1, p_1, p_2 \in [1, 10^4]$.





Optimality of Interpolation Points

Theorem

[BAUR/BEATTIE/B./GUGERCIN 2011]

For special parameterized SISO systems,

$$A(p) \equiv A_0, \quad E(p) \equiv E_0, \quad B(p) = B_0 + p_1 B_1, \quad C(p) = C_0 + p_2 C_1,$$

optimal choice possible, **necessary conditions**:

If \hat{G} minimizes the approximation error w.r.t. $\|G - \hat{G}\|_{\mathcal{H}_2 \times L_2(\Omega)}$, $p \in \Omega \subset \mathbb{R}^d$, and $\Lambda(\hat{A}, \hat{E}) = \{\hat{\lambda}_1, \dots, \hat{\lambda}_r\}$ (all simple), then the **interpolation frequencies** satisfy

$$s_i = -\hat{\lambda}_i, \quad i = 1, \dots, r,$$

and the **parameter interpolation points** $\{p^{(1)}, \dots, p^{(r)}\}$ satisfy the **interpolation conditions**

$$\begin{aligned} G(-\hat{\lambda}_k, p^{(k)}) &= \hat{G}(-\hat{\lambda}, p^{(k)}), \\ \frac{\partial}{\partial s} G(-\hat{\lambda}, p^{(k)}) &= \frac{\partial}{\partial s} \hat{G}(-\hat{\lambda}, p^{(k)}), \quad \nabla_p G(-\hat{\lambda}, p^{(k)}) = \nabla_p \hat{G}(-\hat{\lambda}, p^{(k)}). \end{aligned}$$



Optimality of Interpolation Points

Theorem

[BAUR/BEATTIE/B./GUGERCIN 2011]

For special parameterized SISO systems,

$$A(p) \equiv A_0, \quad E(p) \equiv E_0, \quad B(p) = B_0 + p_1 B_1, \quad C(p) = C_0 + p_2 C_1,$$

optimal choice possible, **necessary conditions**:

If \hat{G} minimizes the approximation error w.r.t. $\|G - \hat{G}\|_{\mathcal{H}_2 \times L_2(\Omega)}$, $p \in \Omega \subset \mathbb{R}^d$, the **parameter interpolation points** $\{p^{(1)}, \dots, p^{(r)}\}$ satisfy the **interpolation conditions**

$$\begin{aligned} G(-\hat{\lambda}_k, p^{(k)}) &= \hat{G}(-\hat{\lambda}, p^{(k)}), \\ \frac{\partial}{\partial s} G(-\hat{\lambda}, p^{(k)}) &= \frac{\partial}{\partial s} \hat{G}(-\hat{\lambda}, p^{(k)}), \quad \nabla_p G(-\hat{\lambda}, p^{(k)}) = \nabla_p \hat{G}(-\hat{\lambda}, p^{(k)}). \end{aligned}$$

Proof:

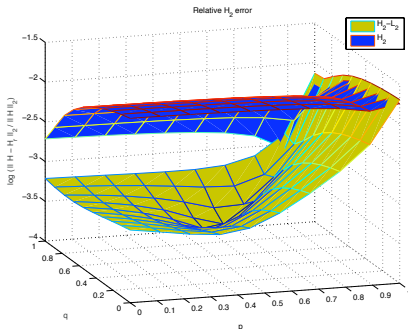
$$\|G\|_{\mathcal{H}_2 \times L_2(\Omega)} = \|L^T \tilde{G} L\|_{\mathcal{H}_2}, \quad \text{where } \tilde{G}(s) = \begin{bmatrix} C_0 \\ C_1 \end{bmatrix} (sE - A)^{-1} [B_0, B_1], \quad L = \begin{bmatrix} 1 & 0 \\ \frac{1}{2} & \frac{1}{2\sqrt{3}} \end{bmatrix}.$$

\Rightarrow **Computation via IRKA applied to \tilde{G} .**



Optimality of Interpolation Points — Numerical Example

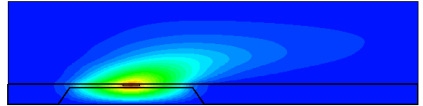
- Model for evolution of temperature distribution on a plate, described by the heat equation.
- FDM SISO model of order $n = 197$.
- Parameter $p_1 \in [0, 1]$ encodes movement of heat source from B_0 to $B_0 + B_1$, analogous for relocation of measurement.



Relative $\mathcal{H}_2 \otimes L_2(\Omega)$ error: 7.5×10^{-4} .



Consider an **anemometer**, a flow sensing device located on a membrane used in the context of minimizing heat dissipation.



Source: [BAUR/B./GREINER/KORVINK/LIENEMANN/MOOSMANN 2011]

- FE model:

$$E\dot{x}(t) = (A + pA_1)x(t) + Bu(t), \quad y(t) = Cx(t), \quad x(0) = 0,$$

- $n = 29,008$, $m = 1$, $q = 3$, $p_1 \in [0, 1]$ fluid velocity.

Source: MOR Wiki: <http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Anemometer>



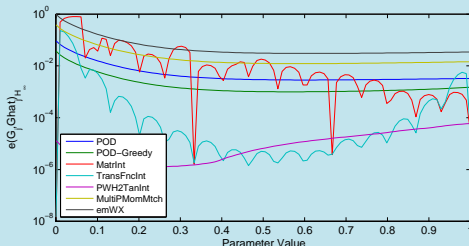
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\mathcal{H}_∞ error



Source: MOR Wiki: <http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Anemometer>



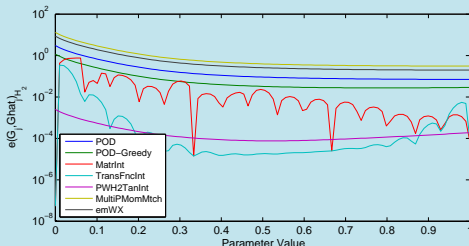
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\mathcal{H}_2 error



Source: MOR Wiki: <http://morwiki.mpi-magdeburg.mpg.de/morwiki/index.php/Anemometer>

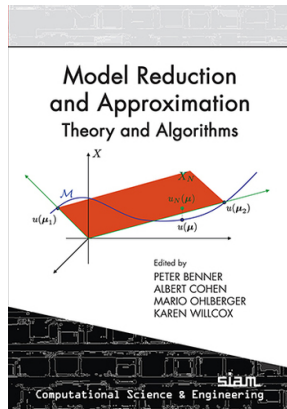


For more details of this comparisons, and other tests, see



U. Baur, P. Benner, B. Haasdonk, C. Himpe, I. Maier, and M. Ohlberger.
Comparison of Methods for Parametric Model Order Reduction of Unsteady Problems.
In P. Benner, A. Cohen, M. Ohlberger, and K. Willcox (eds.), *Model Reduction and Approximation: Theory and Algorithms*.
SIAM, Philadelphia, PA, 2017.

Chapter 9 in





1. Introduction

2. PMOR Methods based on Interpolation

3. PMOR via Bilinearization

Parametric Systems as Bilinear Systems

Balanced Truncation for Bilinear Systems

\mathcal{H}_2 -Model Reduction for Bilinear Systems

Some Notes on LTV/LPV Systems

4. Conclusions and Outlook



Linear Parametric Systems — An Alternative Interpretation

Consider **bilinear control systems**:

$$\Sigma : \quad \begin{cases} \dot{x}(t) = Ax(t) + \sum_{i=1}^m A_i x(t) u_i(t) + Bu(t), \\ y(t) = Cx(t), \quad x(0) = x_0, \end{cases}$$

where $A, A_i \in \mathbb{R}^{n \times n}$, $B \in \mathbb{R}^{n \times m}$, $C \in \mathbb{R}^{q \times n}$.



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where $A, A_i \in \mathbb{R}^{n \times n}$, $B \in \mathbb{R}^{n \times m}$, $C \in \mathbb{R}^{q \times n}$.

Key Observation

[B./Breiten 2011]

Consider parameters as additional inputs, a linear parametric system

$$\dot{x}(t) = Ax(t) + \sum_{i=1}^{m_p} a_i(p) A_i x(t) + B_0 u_0(t), \quad y(t) = Cx(t)$$

with $B_0 \in \mathbb{R}^{n \times m_0}$ can be interpreted as bilinear system:

$$\begin{aligned} u(t) &:= [a_1(p) \quad \dots \quad a_{m_p}(p) \quad u_0(t)]^T, \\ B &:= [\mathbf{0} \quad \dots \quad \mathbf{0} \quad B_0] \in \mathbb{R}^{n \times m}, \quad m = m_p + m_0. \end{aligned}$$



Linear Parametric Systems — An Alternative Interpretation

Linear parametric systems can be interpreted as bilinear systems.



Linear Parametric Systems — An Alternative Interpretation

Linear parametric systems can be interpreted as bilinear systems.

Consequence

Model order reduction techniques for bilinear systems can be applied to linear parametric systems!

Here:

- Balanced truncation,
- \mathcal{H}_2 optimal model reduction.



Balanced Truncation for Linear Systems

Idea (for simplicity, $E = I_n$)

- $\Sigma: \begin{cases} \dot{x}(t) = Ax(t) + Bu(t), \\ y(t) = Cx(t), \end{cases}$ with A stable, i.e., $\Lambda(A) \subset \mathbb{C}^-$,

is **balanced**, if **system Gramians**, i.e., solutions P, Q of the **Lyapunov equations**

$$AP + PA^T + BB^T = 0, \quad A^T Q + QA + C^T C = 0,$$

satisfy: $P = Q = \text{diag}(\sigma_1, \dots, \sigma_n)$ with $\sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_n > 0$.



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- $\{\sigma_1, \dots, \sigma_n\}$ are the Hankel singular values (HSVs) of Σ .
- Compute balanced realization (needs P, Q !) of the system via **state-space transformation**

$$\begin{aligned} \mathcal{T} : (A, B, C) &\mapsto (TAT^{-1}, TB, CT^{-1}) \\ &= \left(\begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix}, \begin{bmatrix} B_1 \\ B_2 \end{bmatrix}, \begin{bmatrix} C_1 & C_2 \end{bmatrix} \right). \end{aligned}$$



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- Truncation $\rightsquigarrow (\hat{A}, \hat{B}, \hat{C}) = (A_{11}, B_1, C_1)$.



Balanced Truncation for Linear Systems

Properties

- Reduced-order model is stable with HSVs $\sigma_1, \dots, \sigma_{\hat{n}}$.



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- Adaptive choice of r via computable error bound:

$$\|y - \hat{y}\|_2 \leq \left(2 \sum_{k=\hat{n}+1}^n \sigma_k \right) \|u\|_2.$$



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Practical implementation

- Rather than solving Lyapunov equations for P, Q (n^2 unknowns!), find $S, R \in \mathbb{R}^{n \times s}$ with $s \ll n$ such that $P \approx SS^T$, $Q \approx RR^T$.
- Reduced-order model directly obtained via small-scale ($s \times s$) SVD of $R^T S$!
- No $\mathcal{O}(n^3)$ or $\mathcal{O}(n^2)$ computations necessary!



The concept of **balanced truncation** can be generalized to the case of bilinear systems, where we need the solutions of the **generalized Lyapunov equations**:

$$AP + PA^T + \sum_{i=1}^m A_i P A_i^T + BB^T = 0,$$
$$A^T Q + QA^T + \sum_{i=1}^m A_i^T Q A_i + C^T C = 0.$$

- These equations also appear for stochastic control systems, see [B./DAMM 2011] and **→ tutorial by T. Damm.**
- "Twice-the-trail-of-the-HSVs" error bound does not hold [B./DAMM 2014].
- Stability preservation [B./DAMM/REDMANN/RODRIGUEZ CRUZ 2016].



Some basic facts and assumptions

$$AX + XA^T + \sum_{i=1}^m A_i X A_i^T + BB^T = 0. \quad (1)$$

- Need a **positive semi-definite symmetric solution X** .



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- In **standard Lyapunov case**, existence and uniqueness guaranteed if A **stable** ($\lambda(A) \subset \mathbb{C}^-$); this is not sufficient here: (1) is equivalent to

$$\left(I_n \otimes A + A \otimes I_n + \sum_{i=1}^m A_i \otimes A_i \right) \text{vec}(X) = -\text{vec}(BB^T).$$

A sufficient condition for stable A is smallness of A_i (related to stability radius of \mathcal{A})
 \rightsquigarrow **bounded-input bounded-output (BIBO) stability** of bilinear systems.

This will be assumed from here on, hence: **existence and uniqueness of positive semi-definite solution $X = X^T$** .



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- Want: solution methods for large scale problems, i.e., only matrix-matrix multiplication with A, A_j , solves with (shifted) A allowed!
- Requires to compute data-sparse approximation to generally dense X ; here:
 $X \approx ZZ^T$ with $Z \in \mathbb{R}^{n \times n_Z}$, $n_Z \ll n$!



Existence of low-rank approximations

Q: Can we expect **low-rank approximations** $ZZ^T \approx X$ to the solution of

$$AX + XA^T + \sum_{j=1}^m A_j X A_j^T + BB^T = 0 ?$$



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Theorem

[B./Breiten 2013]

Assume existence and uniqueness assumption with stable A and $A_j = U_j V_j^T$, with $U_j, V_j \in \mathbb{R}^{n \times r_j}$. Set $r = \sum_{j=1}^m r_j$.
Then the solution X of

$$AX + XA^T + \sum_{j=1}^m A_j X A_j^T + BB^T = 0$$

can be approximated by X_k of **rank** $(2k+1)(m+r)$, with an error satisfying

$$\|X - X_k\|_2 \lesssim \exp(-\sqrt{k}).$$



Numerical Methods

- Generalized Alternating Directions Iteration (ADI) method.
 1. Computing square solution matrix ($\sim n^2$ unknowns) [DAMM 2008].
 2. Computing low-rank factors of solutions ($\sim n$ unknowns) [B./BREITEN 2013].
- Generalized Extended (or rational) Krylov Subspace Method (EKSM) [B./BREITEN 2013].
- Tensorized versions of standard Krylov subspace methods, e.g., PCG, PBiCGStab [KRESSNER/TOBLER 2011, B./BREITEN 2013].
- Inexact stationary (fix point) iteration [SHANK/SIMONCINI/SZYLD 2016].



Fast Simulation of Cyclic Voltammograms [FENG/KOZIOL/RUDNYI/KORVINK 2006]

$$E\dot{x}(t) = (A + p_1(t)A_1 + p_2(t)A_2)x(t) + B,$$
$$y(t) = Cx(t), \quad x(0) = x_0 \neq 0,$$

- Rewrite as system with zero initial condition,
- FE model: $n = 16,912$, $m = 3$, $q = 1$,
- $p_j \in [0, 10^9]$ time-varying voltage functions,
- transfer function $G(i\omega, p_1, p_2)$,
- reduced system dimension $r = 67$,
- $$\max_{\substack{\omega \in \{\omega_{\min}, \dots, \omega_{\max}\} \\ p_j \in \{p_{\min}, \dots, p_{\max}\}}} \frac{\|G - \hat{G}\|_2}{\|G\|_2} < 6 \cdot 10^{-4},$$
- evaluation times: FOM 4.5h, ROM 38s
 \rightsquigarrow speed-up factor ≈ 426 .

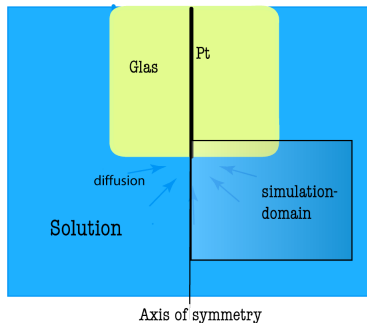


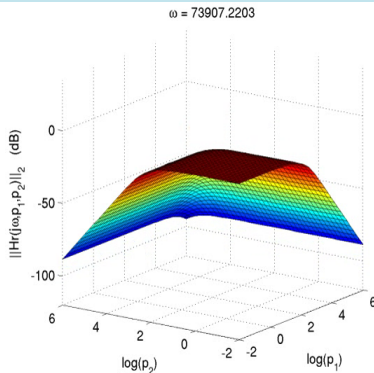
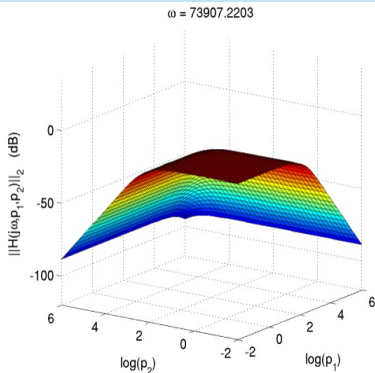
Figure : [FENG ET AL. 2006]



Fast Simulation of Cyclic Voltammogramms [FENG/KOZIOL/RUDNYI/KORVINK 2006]

Original...

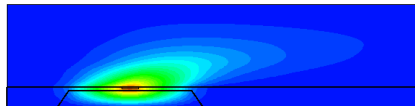
and reduced-order model.





2D Model of an Anemometer [BAUR ET AL. 2011]

Consider again the **anemometer** example.



$$E\dot{x}(t) = (A + pA_1)x(t) + Bu(t), \quad y(t) = Cx(t), \quad x(0) = 0,$$

- FE model: $n = 29,008$, $m = 1$, $q = 3$,
- $p_1 \in [0, 1]$ fluid velocity,
- transfer function $G(i\omega, p_1)$, reduced system dimension $r = 146$,
- $$\max_{\substack{\omega \in \{\omega_{min}, \dots, \omega_{max}\} \\ p_1 \in \{p_{min}, \dots, p_{max}\}}} \frac{\|G(\omega, p) - \hat{G}(\omega, p)\|_2}{\|G(\omega, p)\|_2} < 3 \cdot 10^{-5},$$
- evaluation times: **FOM 51min**, **ROM 21sec**.



Some background

Consider bilinear system ($m = 1$, i.e. SISO)

$$\Sigma : \left\{ \begin{array}{l} \dot{x}(t) = Ax(t) + A_1x(t)u(t) + Bu(t), \\ y(t) = Cx(t). \end{array} \right.$$



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Output Characterization (SISO) via Volterra series:

$$y(t) = \sum_{k=1}^{\infty} \int_0^t \int_0^{t_1} \cdots \int_0^{t_{k-1}} K(t_1, \dots, t_k) u(t - t_1 - \dots - t_k) \cdots u(t - t_k) dt_k \cdots dt_1,$$

with kernels $K(t_1, \dots, t_k) = Ce^{At_k} A_1 \cdots e^{At_2} A_1 e^{At_1} B$.



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Multivariate Laplace-transform:

$$G_k(s_1, \dots, s_k) = C(s_k I - A)^{-1} A_1 \cdots (s_2 I - A)^{-1} A_1 (s_1 I - A)^{-1} B.$$



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Bilinear \mathcal{H}_2 -norm:

[ZHANG/LAM 2002]

$$\|\Sigma\|_{\mathcal{H}_2} := \left(\text{tr} \left(\left(\sum_{k=1}^{\infty} \int_{-\infty}^{\infty} \cdots \int_{-\infty}^{\infty} \frac{1}{(2\pi)^k} \overline{G_k(i\omega_1, \dots, i\omega_k)} G_k^T(i\omega_1, \dots, i\omega_k) \right) \right) \right)^{\frac{1}{2}}.$$



Measuring the Approximation Error

Lemma

[B./Breiten 2012]

Let Σ denote a bilinear system. Then, the \mathcal{H}_2 -norm is given as:

$$\|\Sigma\|_{\mathcal{H}_2}^2 = (\text{vec}(I_q))^T (C \otimes C) \left(-A \otimes I - I \otimes A - \sum_{i=1}^m A_i \otimes A_i \right)^{-1} (B \otimes B) \text{vec}(I_m).$$



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Error System

In order to find an \mathcal{H}_2 -optimal reduced system, define the **error system** $\Sigma^{err} := \Sigma - \hat{\Sigma}$ as follows:

$$A^{err} = \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix}, \quad A_i^{err} = \begin{bmatrix} A_i & 0 \\ 0 & \hat{A}_i \end{bmatrix}, \quad B^{err} = \begin{bmatrix} B \\ \hat{B} \end{bmatrix}, \quad C^{err} = [C \quad -\hat{C}].$$

**Theorem (error system squared)**

Let Σ and $\hat{\Sigma}$ be the original and reduced bilinear systems, respectively. Then

$$\begin{aligned}
 \mathcal{E} &:= \|\Sigma^{err}\|_{\mathcal{H}_2}^2 := \|\Sigma - \hat{\Sigma}\|_{\mathcal{H}_2}^2 \\
 &= (\mathcal{I}_p)^T ([C \quad -\tilde{C}] \otimes [C \quad -\hat{C}]) \times \\
 &\quad \left(- \begin{bmatrix} A & 0 \\ 0 & \Lambda \end{bmatrix} \otimes I_{n+r} - I_{n+r} \otimes \begin{bmatrix} A & 0 \\ 0 & \hat{A} \end{bmatrix} - \sum_{k=1}^m \begin{bmatrix} A_k & 0 \\ 0 & \tilde{A}_k^T \end{bmatrix} \otimes \begin{bmatrix} A_k & 0 \\ 0 & \hat{A}_k \end{bmatrix} \right)^{-1} \times \\
 &\quad \left(\begin{bmatrix} B \\ \tilde{B}^T \end{bmatrix} \otimes \begin{bmatrix} B \\ \hat{B} \end{bmatrix} \right) \mathcal{I}_m, \\
 &= (\mathcal{I}_p)^T ([C \quad -\tilde{C}] \otimes [C \quad -\tilde{C}]) \times \\
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 \end{aligned}$$

where $\mathcal{I}_\ell = \text{vec}(I_\ell)$, $R\Lambda R^{-1} = \hat{A}$ is the spectral decomposition of \hat{A} ,
and $\tilde{B} = \hat{B}^T R^{-T}$, $\tilde{C} = \hat{C}R$, $\tilde{A}_k = R^T \hat{A}_k^T R^{-T}$.



\mathcal{H}_2 -Optimality Conditions

Assume $\hat{\Sigma}$ is given in coordinate system induced by **eigendecomposition** of \hat{A} :

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Using Λ , \tilde{A}_i , \tilde{B} , \tilde{C} as optimization parameters, we can derive **necessary conditions for \mathcal{H}_2 -optimality**, e.g., $\frac{\partial \mathcal{E}}{\partial \tilde{C}_{ij}} = 0$ yields:

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$$\begin{aligned} & (\text{vec}(I_q))^T \left(e_j e_\ell^T \otimes C \right) \left(-\Lambda \otimes I_n - I_r \otimes A - \sum_{k=1}^m \tilde{A}_k \otimes A_k \right)^{-1} \left(\tilde{B} \otimes B \right) \text{vec}(I_m) \\ &= (\text{vec}(I_q))^T \left(e_j e_\ell^T \otimes \hat{C} \right) \left(-\Lambda \otimes I_n - I_r \otimes \hat{A} - \sum_{k=1}^m \tilde{A}_k \otimes \hat{A}_k \right)^{-1} \left(\tilde{B} \otimes \hat{B} \right) \text{vec}(I_m). \end{aligned}$$



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Connection to interpolation of transfer functions?



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For $A_k \equiv 0$, this is equivalent to

$$G(-\lambda_\ell) \tilde{B}_\ell^T = \hat{G}(-\lambda_\ell) \tilde{B}_\ell^T, \quad \ell = 1, \dots, r.$$

\rightsquigarrow Tangential interpolation at mirror images of reduced system poles!

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Note: [FLAGG 2011] shows equivalence to interpolating the Volterra series!



Algorithm 3 Bilinear IRKA

Input: $A, A_i, B, C, \hat{A}, \hat{A}_i, \hat{B}, \hat{C}$

Output: $A^{opt}, A_i^{opt}, B^{opt}, C^{opt}$

1: **while** (change in $\Lambda > \epsilon$) **do**

2: $R\Lambda R^{-1} = \hat{A}, \tilde{B} = R^{-1}\hat{B}, \tilde{C} = \hat{C}R, \tilde{A}_i = R^{-1}\hat{A}_iR$

3: $\text{vec}(V) = \left(-\Lambda \otimes I_n - I_r \otimes A - \sum_{i=1}^m \tilde{A}_i \otimes A_i \right)^{-1} (\tilde{B} \otimes B) \text{vec}(I_m)$

4: $\text{vec}(W) = \left(-\Lambda \otimes I_n - I_r \otimes A^T - \sum_{i=1}^m \tilde{A}_i^T \otimes A_i^T \right)^{-1} (\tilde{C}^T \otimes C^T) \text{vec}(I_q)$

5: $V = \text{orth}(V), W = \text{orth}(W)$

6: $\hat{A} = (W^T V)^{-1} W^T A V, \hat{A}_i = (W^T V)^{-1} W^T A_i V,$
 $\hat{B} = (W^T V)^{-1} W^T B, \hat{C} = C V$

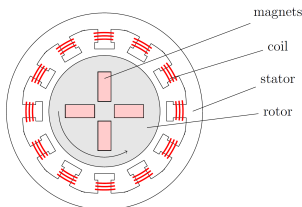
7: **end while**

8: $A^{opt} = \hat{A}, A_i^{opt} = \hat{A}_i, B^{opt} = \hat{B}, C^{opt} = \hat{C}$



Industrial Case Study: Thermal Analysis of Electrical Motor

- Thermal simulations to detect whether temperature changes lead to fatigue or deterioration of employed materials.
- Main heat source: thermal losses resulting from current stator coil/rotor.
- Many different current profiles need to be considered to predict whether temperature on certain parts of the motor remains in feasible region.
- Finite element analysis on rather complicated geometries \rightsquigarrow large-scale linear models with 7/13 parameters.



Schematic view of an electrical motor.



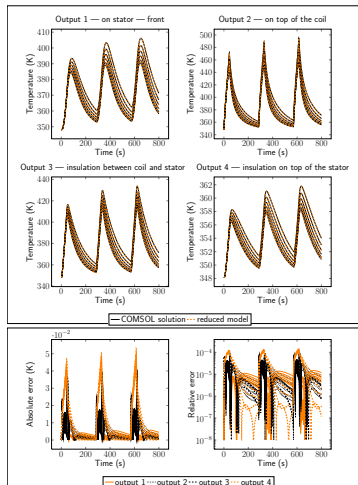
Bosch integrated motor generator used in hybrid variants of Porsche Cayenne, VW Touareg.

Pictures:  **BOSCH**



Industrial Case Study: Thermal Analysis of Electrical Motor

- FEM analysis of thermal model \rightsquigarrow linear parametric systems with $n = 41,199$, $m = 4$ inputs, and $d = 13$ parameters,
- measurements taken at $q = 4$ heat sensors;
- time for 1 transient simulation in COMSOL[®] $\sim 90\text{min}$;
- ROM order $\hat{n} = 300$, time for 1 transient simulation $\sim 15\text{sec}$.
- Legend: Temperature curves for six different values (5, 25, 45, 65, 85, $100[W/m^2K]$) of the heat transfer coefficient on the coil.



**Linear time-varying (LTV) systems**

$$\begin{aligned} E(t)\dot{x}(t) &= A(t)x(t) + B(t)u(t), & E(t), A(t) &\in \mathbb{R}^{n \times n}, \\ y(t) &= C(t)x(t), & B(t) &\in \mathbb{R}^{n \times m}, C(t) \in \mathbb{R}^{q \times n}, \end{aligned}$$

Assume the matrix functions $A(t), \dots$ to be sufficiently smooth.



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Reduced LTV systems

Goal: efficiently compute reduced-order model (ROM):

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- Balanced truncation was generalized to LTV systems [Shokoochi/Silverman/Van Dooren 1983, Lang/Saak/Stykel 2016], but requires time-dependent projection matrices (computed by [Lyapunov differential eqns](#)) \rightsquigarrow **evaluating ROM needs resorting to $\mathcal{O}(n)$ computations; hence, MOR target missed!**

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- **Can apply MOR methods for bilinear systems!**



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Model order reduction techniques for bilinear systems can be applied to LTV/LPV systems!

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Outlook:

- Not implemented yet.
- Need practical examples.
- Error bounds???



1. Introduction
2. PMOR Methods based on Interpolation
3. PMOR via Bilinearization
4. Conclusions and Outlook



- We have reviewed some of the most popular PMOR methods developed in the last decade, in particular those based on rational interpolation.

Open problem in general: optimal interpolation points.



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- Most of the methods can be used to significantly accelerate UQ by Monte Carlo or Stochastic Collocation methods!



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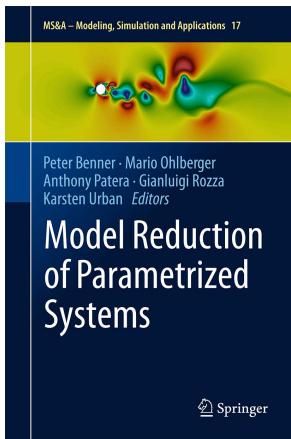


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Proceedings of MoRePaS III out soon ...



MoRePaS IV — Nantes, April 10–13, 2018.

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